

$$k_9 = \frac{2.0 \sqrt{a b}}{a + b} \tag{4.4.2.11}$$

$$k_8 = \frac{k_9}{\sqrt{1.0 - \frac{(b - a)^2}{(2.0 c)^2}}} \tag{4.4.2.12}$$

These equations assume loose coupling and that the substrate is infinitely thick.

REFERENCES

- [1] Bandrand, H., *et al.*, "Bias-Variable Characteristics of Coupled Coplanar Waveguide on GaAs Substrate," *Electronics Letters*, Vol. 23, No. 4, February 13, 1987, pp. 171–172.
- [2] Chang, Ching Ten, and Graham A. Garcia, "Crosstalk Between Two Coplanar Waveguides," *Archiv Für Elektronik und Übertragungstechnik*, Band 43, Heft 1, 1989, pp. 55–58.
- [3] Ghione, Giovanni, and Carlo U. Naldi, "Coplanar Waveguides for MMIC Applications: Effect of Upper Shielding Conductor Backing, Finite-Extent Ground Planes, and Line-to-Line Coupling," *IEEE Transactions on Microwave Theory and Techniques*, Vol. MTT-35, No. 3, March 1987, pp. 260–267.
- [4] Wen, Cheng P., "Coplanar-Waveguide Directional Couplers," *IEEE Transactions on Microwave Theory and Techniques*, Vol. MTT-18, No. 6, June 1970, pp. 318–322.

4.4.3 Edge-Coupled CPWG

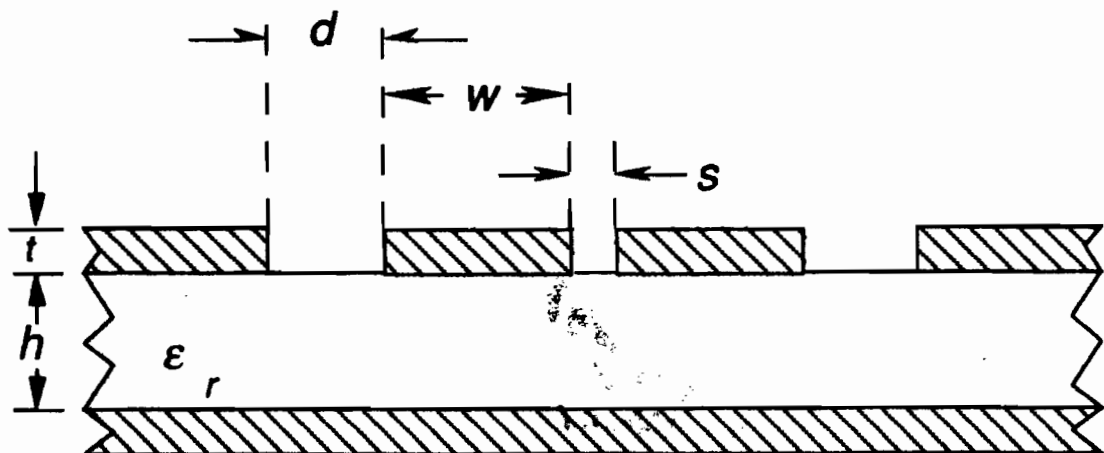


Figure 4.4.3.1: Edge-Coupled CPWG

$$Z_{0,o} = \frac{\eta_0}{\sqrt{\epsilon_{eff,o}}} \left[ \frac{1.0}{2.0 \frac{K(k_o)}{K'(k_o)} + \frac{K(\beta_1)}{K'(\beta_1)}} \right] (\Omega) \quad (4.4.3.1)$$

$$Z_{0,e} = \frac{\eta_0}{\sqrt{\epsilon_{eff,e}}} \left[ \frac{1.0}{2.0 \frac{K(k_e)}{K'(k_e)} + \frac{K(\beta_1 k_1)}{K'(\beta_1 k_1)}} \right] (\Omega) \quad (4.4.3.2)$$

$$\epsilon_{eff,o} = \frac{2.0 \epsilon_r \frac{K(k_o)}{K'(k_o)} + \frac{K(\beta_1)}{K'(\beta_1)}}{2.0 \frac{K(k_o)}{K'(k_o)} + \frac{K(\beta_1)}{K'(\beta_1)}} \quad (4.4.3.3)$$

$$\epsilon_{eff,e} = \frac{2.0 \epsilon_r \frac{K(k_e)}{K'(k_e)} + \frac{K(\beta_1 k_1)}{K'(\beta_1 k_1)}}{2.0 \frac{K(k_e)}{K'(k_e)} + \frac{K(\beta_1 k_1)}{K'(\beta_1 k_1)}} \quad (4.4.3.4)$$

where

$$k_o = \Lambda \frac{-\sqrt{\Lambda^2 - t_c^2} + \sqrt{\Lambda^2 - t_B^2}}{t_B \sqrt{\Lambda^2 - t_c^2} + t_c \sqrt{\Lambda^2 - t_B^2}} \quad (4.4.3.5)$$

$$k_e = \Lambda' \frac{-\sqrt{\Lambda'^2 - t'_c{}^2} + \sqrt{\Lambda'^2 - t'_B{}^2}}{t'_B \sqrt{\Lambda'^2 - t'_c{}^2} + t'_c \sqrt{\Lambda'^2 - t'_B{}^2}} \quad (4.4.3.6)$$

$$\Lambda = \frac{\sinh^2 \left[ \frac{\pi (s / 2.0 + w + d)}{2.0 h} \right]}{2} \quad (4.4.3.7)$$

$$t_c = \sinh^2 \left[ \frac{\pi (s / 2.0 + w)}{2.0 h} \right] - \Lambda \quad (4.4.3.8)$$

$$t_B = \sinh^2 \left( \frac{\pi s}{4.0 h} \right) - \Lambda \quad (4.4.3.9)$$

$$\Lambda' = \frac{\cosh^2 \left[ \frac{\pi (s / 2.0 + w + d)}{2.0 h} \right]}{2.0} \quad (4.4.3.10)$$

$$t'_c = \sinh^2 \left[ \frac{\pi (s / 2.0 + w)}{2.0 h} \right] - \Lambda' + 1.0 \quad (4.4.3.11)$$

$$t'_B = \sinh^2 \left[ \frac{\pi s}{4.0 h} \right] - \Lambda' + 1.0 \quad (4.4.3.12)$$

$$\beta_1 = \sqrt{\frac{1.0 - y^2}{1.0 - k_1^2 y^2}} \quad (4.4.3.13)$$

$$y = \frac{s}{s + 2.0 w} \quad (4.4.3.14)$$

$$k_1 = \frac{s + 2.0 w}{s + 2.0 w + 2.0 d} \quad (4.4.3.15)$$

To guarantee coplanar propagation,

$$s + 2.0 w + 2.0 d \leq h.$$

The analysis is a conformal mapping technique, so the equations are exact for low frequency calculation.

#### REFERENCES

- [1] Chang, Ching Ten, and Graham A. Garcia, "Crosstalk Between Two Coplanar Waveguides," *Archiv Für Elektronik und Übertragungstechnik*, Band 43, Heft 1, 1989, pp. 55–58. (Edge-coupled CPW's separated by ground plane.)
- [2] Hanna, Victor Fouad, "Parameters of Coplanar Directional Couplers with Lower Ground Plane," *15th European Microwave Conference Proceedings*, 1985, pp. 820–825. ([2] dropped primes in (8), corrected here.)